

6.003: Signals and Systems

Frequency Response

March 4, 2010

Review

Last time, we saw how a linear, time-invariant (LTI) system can be characterized by its unit-sample/impulse response.

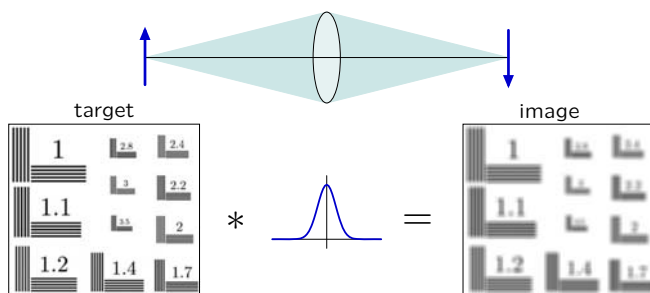
$$\text{DT: } y[n] = (x * h)[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$

$$\text{CT: } y(t) = (x * h)(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau)d\tau$$

Characterizing a system by its unit-sample/impulse response is especially insightful for some systems.

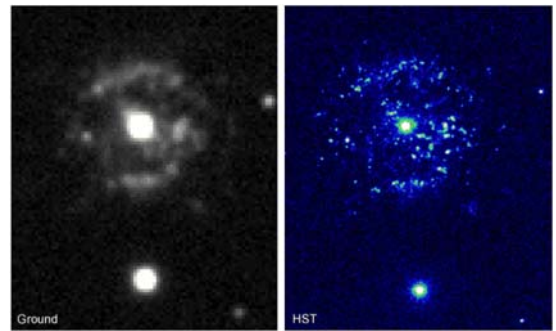
Microscope

Blurring can be represented by convolving the image with the optical "point-spread-function" (3D impulse response).



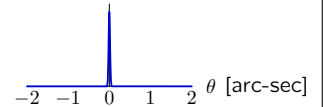
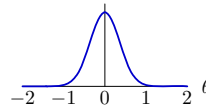
Blurring is inversely related to the diameter of the lens.

Hubble Space Telescope



optical + atmospheric blurring

optical blurring



Frequency Response

Today we will investigate a different way to characterize a system: the **frequency response**.

Many systems are naturally described by their responses to sinusoids.

Example: audio systems

Check Yourself

How were frequencies modified in following music clips?

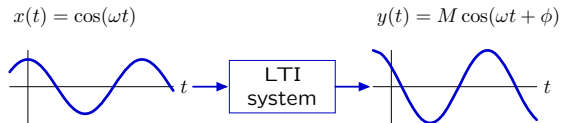
HF: high frequencies ↑: increased
 LF: low frequencies ↓: decreased

- | | clip 1 | clip 2 |
|----|-------------------|--------|
| 1. | HF↑ | HF↓ |
| 2. | LF↑ | LF↓ |
| 3. | HF↑ | LF↓ |
| 4. | LF↑ | HF↓ |
| 5. | none of the above | |

Frequency Response Preview

If the input to a linear, time-invariant system is an eternal sinusoid, then the output is also an eternal sinusoid:

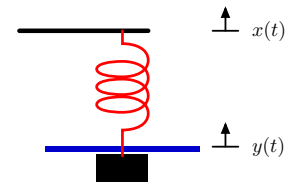
- same frequency
- possibly different amplitude, and
- possibly different phase angle.



The **frequency response** is a plot of the magnitude M and angle ϕ as a function of frequency ω .

Demonstration

Measure the frequency response of a mass, spring, dashpot system.

**Frequency Response**

Calculate the frequency response.

Methods

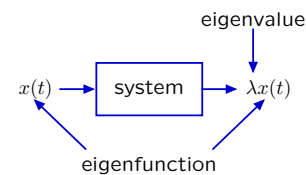
- solve differential equation
 - find particular solution for $x(t) = \cos \omega_0 t$
- find impulse response of system
 - convolve with $x(t) = \cos \omega_0 t$

New method

- use eigenfunctions and eigenvalues

Eigenfunctions

If the output signal is a scalar multiple of the input signal, we refer to the signal as an eigenfunction and the multiplier as the eigenvalue.

**Check Yourself: Eigenfunctions**

Consider the system described by

$$\dot{y}(t) + 2y(t) = x(t).$$

Determine if each of the following functions is an eigenfunction of this system. If it is, find its eigenvalue.

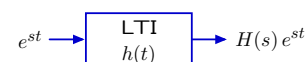
1. e^{-t} for all time
2. e^t for all time
3. e^{jt} for all time
4. $\cos(t)$ for all time
5. $u(t)$ for all time

Complex Exponentials

Complex exponentials are eigenfunctions of LTI systems.

If $x(t) = e^{st}$ and $h(t)$ is the impulse response then

$$y(t) = (h * x)(t) = \int_{-\infty}^{\infty} h(\tau) e^{s(t-\tau)} d\tau = e^{st} \int_{-\infty}^{\infty} h(\tau) e^{-s\tau} d\tau = H(s) e^{st}$$



Eternal sinusoids are sums of complex exponentials.

$$\cos \omega_0 t = \frac{1}{2} (e^{j\omega_0 t} + e^{-j\omega_0 t})$$

Furthermore, the eigenvalue associated with e^{st} is $H(s)$!

Rational System Functions

Eigenvalues are particularly easy to evaluate for systems represented by linear differential equations with constant coefficients.

Then the system function is a ratio of polynomials in s .

Example:

$$\ddot{y}(t) + 3\dot{y}(t) + 4y(t) = 2\dot{x}(t) + 7x(t) + 8x(t)$$

Then

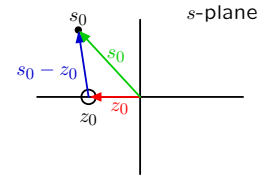
$$H(s) = \frac{2s^2 + 7s + 8}{s^2 + 3s + 4} \equiv \frac{N(s)}{D(s)}$$

Vector Diagrams

The value of $H(s)$ at a point $s = s_0$ can be determined graphically using vectorial analysis.

Factor the numerator and denominator of the system function to make poles and zeros explicit.

$$H(s_0) = K \frac{(s_0 - z_0)(s_0 - z_1)(s_0 - z_2) \cdots}{(s_0 - p_0)(s_0 - p_1)(s_0 - p_2) \cdots}$$



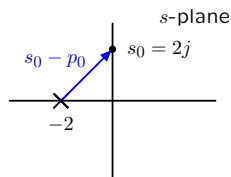
Each factor in the numerator/denominator corresponds to a vector from a zero/pole (here z_0) to s_0 , the point of interest in the s -plane.

Vector Diagrams

Example: Find the response of the system described by

$$H(s) = \frac{1}{s + 2}$$

to the input $x(t) = e^{2jt}$ (for all time).



The denominator of $H(s)|_{s=2j}$ is $2j + 2$, a vector with length $2\sqrt{2}$ and angle $\pi/4$. Therefore, the response of the system is

$$y(t) = H(2j)e^{2jt} = \frac{1}{2\sqrt{2}}e^{-j\frac{\pi}{4}}e^{2jt}$$

Vector Diagrams

The value of $H(s)$ at a point $s = s_0$ can be determined by combining the contributions of the vectors associated with each of the poles and zeros.

$$H(s_0) = K \frac{(s_0 - z_0)(s_0 - z_1)(s_0 - z_2) \cdots}{(s_0 - p_0)(s_0 - p_1)(s_0 - p_2) \cdots}$$

The magnitude is determined by the product of the magnitudes.

$$|H(s_0)| = |K| \frac{|(s_0 - z_0)||s_0 - z_1||s_0 - z_2| \cdots}{|(s_0 - p_0)||s_0 - p_1||s_0 - p_2| \cdots}$$

The angle is determined by the sum of the angles.

$$\angle H(s_0) = \angle K + \angle(s_0 - z_0) + \angle(s_0 - z_1) + \cdots - \angle(s_0 - p_0) - \angle(s_0 - p_1) - \cdots$$

Frequency Response

Response to eternal sinusoids.

Let $x(t) = \cos \omega_0 t$ (for all time). Then

$$x(t) = \frac{1}{2} (e^{j\omega_0 t} + e^{-j\omega_0 t})$$

and the response to a sum is the sum of the responses.

$$y(t) = \frac{1}{2} (H(j\omega_0) e^{j\omega_0 t} + H(-j\omega_0) e^{-j\omega_0 t})$$

Conjugate Symmetry

The complex conjugate of $H(j\omega)$ is $H(-j\omega)$.

The system function is the Laplace transform of the impulse response:

$$H(s) = \int_{-\infty}^{\infty} h(t)e^{-st} dt$$

where $h(t)$ is a real-valued function of t for physical systems.

$$H(j\omega) = \int_{-\infty}^{\infty} h(t)e^{-j\omega t} dt$$

$$H(-j\omega) = \int_{-\infty}^{\infty} h(t)e^{j\omega t} dt \equiv (H(j\omega))^*$$

Frequency Response

Response to eternal sinusoids.

Let $x(t) = \cos \omega_0 t$ (for all time), which can be written as

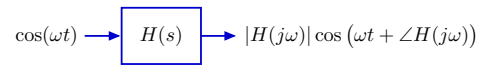
$$x(t) = \frac{1}{2} (e^{j\omega_0 t} + e^{-j\omega_0 t})$$

The response to a sum is the sum of the responses,

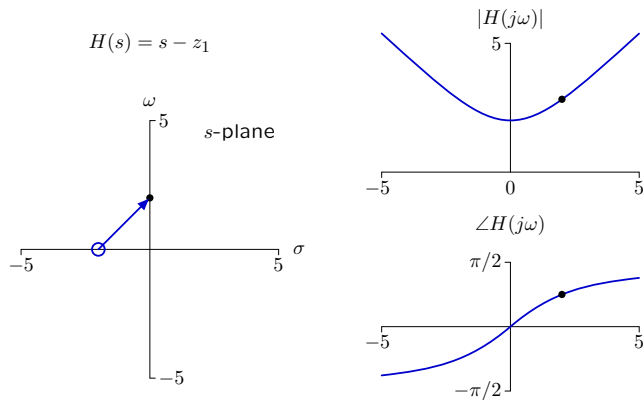
$$\begin{aligned} y(t) &= \frac{1}{2} (H(j\omega_0)e^{j\omega_0 t} + H(-j\omega_0)e^{-j\omega_0 t}) \\ &= \text{Re} \{ H(j\omega_0)e^{j\omega_0 t} \} \\ &= \text{Re} \{ |H(j\omega_0)| e^{j\angle H(j\omega_0)} e^{j\omega_0 t} \} \\ &= |H(j\omega_0)| \text{Re} \{ e^{j\omega_0 t + j\angle H(j\omega_0)} \} \\ y(t) &= |H(j\omega_0)| \cos(\omega_0 t + \angle(H(j\omega_0))). \end{aligned}$$

Frequency Response

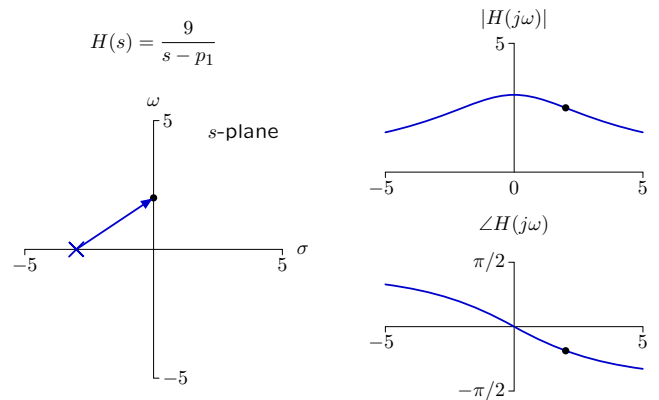
The magnitude and phase of the response of a system to an eternal cosine signal is the magnitude and phase of the system function evaluated at $s = j\omega$.



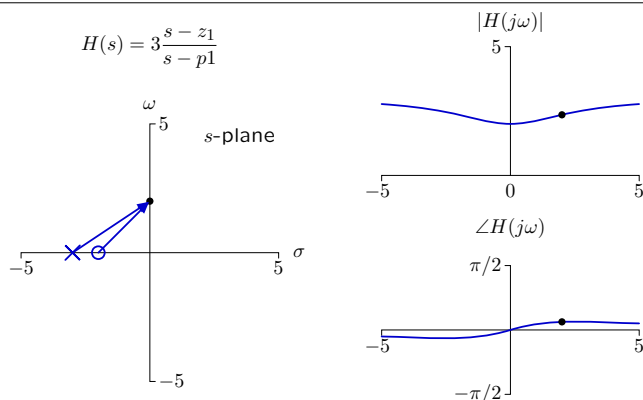
Vector Diagrams



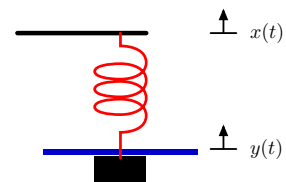
Vector Diagrams



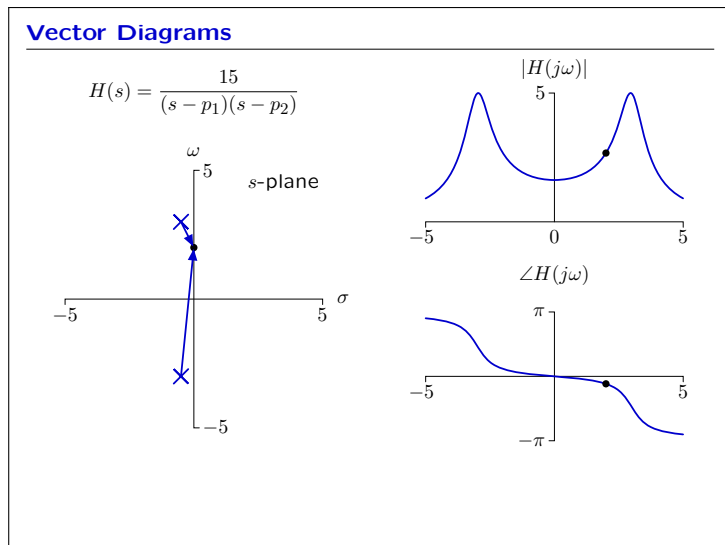
Vector Diagrams



Example: Mass, Spring, and Dashpot



$$\begin{aligned} F &= Ma = M\ddot{y}(t) = K(x(t) - y(t)) - B\dot{y}(t) \\ M\ddot{y}(t) + B\dot{y}(t) + Ky(t) &= Kx(t) \\ (s^2M + sB + K) Y(s) &= KX(s) \\ H(s) &= \frac{K}{s^2M + sB + K} \end{aligned}$$



Check Yourself

Consider the system represented by the following poles.

Find the frequency ω at which the magnitude of the response $y(t)$ is greatest if $x(t) = \cos \omega t$.

1. $\omega = \omega_d$
2. $\omega_d < \omega < \omega_0$
3. $0 < \omega < \omega_d$
4. none of the above

Check Yourself

Consider the system represented by the following poles.

Find the frequency ω at which the phase of the response $y(t)$ is $-\pi/2$ if $x(t) = \cos \omega t$.

0. $0 < \omega < \omega_d$
1. $\omega = \omega_d$
2. $\omega_d < \omega < \omega_0$
3. $\omega = \omega_0$
4. $\omega > \omega_0$
5. none

Frequency Response: Summary

LTI systems can be characterized by responses to eternal sinusoids.

Many systems are naturally described by their frequency response.

- audio systems
- mass, spring, dashpot system

Frequency response is easy to calculate from the system function.

Frequency response lives on the $j\omega$ axis of the Laplace transform.

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